

Initial and Final Value Theorems

Final Value Theorem - determines the steady-state value of the system response without finding the inverse transform.

- Procedure:
- 1.) find the transfer function $X(s)$
 - 2.) multiply $X(s)$ by s
 - 3.) take the limit of $sX(s)$ as S goes to zero
 - 4.) result is value of $x(t)$ when $t =$ infinity

$$\lim_{t \rightarrow \infty} x(t) = \lim_{s \rightarrow 0} s \cdot X(s)$$

Initial Value Theorem - determines the value of the time function when $t=0$ without finding the inverse transform

- Procedure:
- 1.) find the transfer function $X(s)$
 - 2.) multiply $X(s)$ by s
 - 3.) take the limit of $sX(s)$ as S goes to infinity
 - 4.) result is value of $x(t)$ when $t =0$

$$\lim_{t \rightarrow 0} x(t) = \lim_{s \rightarrow \infty} s \cdot X(s)$$

Example 1: Find the initial value of the transfer function

$$X(s) = \frac{808}{s \cdot (s^2 + 2s + 101)}$$

$$X(0) = \lim_{t \rightarrow 0} x(t) = \lim_{s \rightarrow \infty} sX(s) = \lim_{s \rightarrow \infty} \frac{s \cdot 808}{s(s^2 + 2s + 101)}$$

$$\lim_{s \rightarrow \infty} = \frac{808}{s^2 + 2s + 101} \quad \text{Larger values of } s \text{ make denominator larger ratio} \rightarrow 0$$

$$X(0) = \lim_{s \rightarrow \infty} \frac{808}{s^2 + 2s + 101} = \underline{\underline{0}} \quad \text{ANS}$$

Example 2: Find the final value of the transfer function $X(s)$ above.

$$\text{Final Value} \quad \lim_{t \rightarrow \infty} x(t) = \lim_{s \rightarrow 0} sX(s)$$

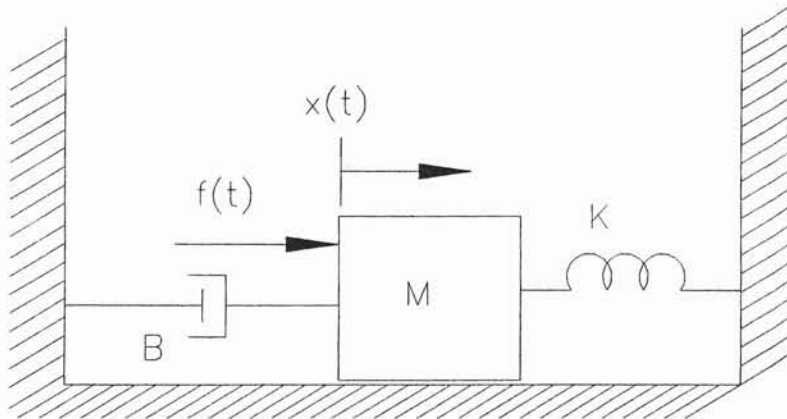
$$\lim_{s \rightarrow 0} sX(s) = \lim_{s \rightarrow 0} \frac{s(808)}{s(s^2 + 2s + 101)} = \lim_{s \rightarrow 0} \frac{808}{s^2 + 2s + 101}$$

$$\text{as } s \rightarrow 0 \quad s^2 \rightarrow 0 \text{ and } 2s \rightarrow 0$$

$$\lim_{s \rightarrow 0} \frac{808}{s^2 + 2s + 101} = \frac{808}{101} = \underline{\underline{8}}$$

$$\text{The final value} \quad \underline{\underline{X(\infty) = 8}}$$

Using Laplace for solving mechanical systems



Write the differential equations for the above system with respect to position and solve them using Laplace transform methods. Assume $f(t) = F$ and that the mass slides on a frictionless surface. $x(0)=0$

$$f(t) = M \cdot \frac{d^2}{dt^2} x(t) + B \cdot \frac{d}{dt} x(t) + K \cdot x(t)$$

$$F = M \cdot \frac{d^2}{dt^2} x(t) + B \cdot \frac{d}{dt} x(t) + K \cdot x(t)$$

Take Laplace transform of both sides

$$\frac{F}{s} = M \cdot s^2 \cdot X(s) + B \cdot s \cdot X(s) + K \cdot X(s)$$

Solve for the position $X(s)$

$$\frac{F}{s(M \cdot s^2 + B \cdot s + K)} = X(s)$$

Let $M = 1$, $F = 5$, $B = 4$ and $K = 5$. Solve this using Laplace and partial fraction expansion.

$$\frac{5}{s(1 \cdot s^2 + 4 \cdot s + 5)} = X(s) \quad \text{Use quadratic formula to factor denominator}$$

$$a = 1 \quad b = 4 \quad c = 5$$

$$s_1 = \frac{-b + \sqrt{b^2 - 4 \cdot a \cdot c}}{2 \cdot a} \quad s_2 = \frac{-b - \sqrt{b^2 - 4 \cdot a \cdot c}}{2 \cdot a}$$

$$s_1 = -2 + i \quad s_2 = -2 - i$$

Factored form of function

$$\frac{5}{s \cdot (s + (2 + j)) \cdot (s + (2 - j))} = X(s)$$

Use partial fraction expansion

$$\frac{5}{s(s + (2 + j))(s + (2 - j))} = \frac{A}{s} + \frac{B}{s + (2 + j)} + \frac{C}{s + (2 - j)}$$

Find A multiply by s

$$\frac{5s}{s(s + (2 + j))(s + (2 - j))} \Big|_{s=0} = \frac{A\cancel{s}}{\cancel{s}} + \frac{Bs}{s + (2 + j)} + \frac{Cs}{s + (2 - j)} \Big|_{s=0}$$

$$\frac{5}{(2 + j)(2 - j)} = A = \frac{5}{2^2 + 1} = \frac{1}{\frac{5}{5}}$$

Mechanical system solution: Continued

Solve For B

$$\frac{5(s+(2+j))}{s(s+(2+j))(s+(2-j))} = \frac{A(s+2+j)}{s} + \frac{B(s+(2+j))}{s+(2+j)} + \frac{C(s+(2+j))}{s+(2-j)}$$

$$\frac{5}{(-2-j)(-2-j+2-j)} = \frac{A(-2+j+2+j)}{s} + B + \frac{C(-2+j+2+j)}{s+(2-j)} \quad s = -(2+j)$$

$$\frac{5}{(-2-j)(-2-j)} = B = \frac{5}{4j-2}$$

Solve for C

$$\frac{5(s+(2-j))}{s(s+(2+j))(s+(2-j))} = \frac{A(s+(2-j))}{s} + \frac{B(s+(2-j))}{s+(2+j)} + \frac{C(s+(2-j))}{s+(2-j)}$$

$$\frac{5}{s(s+(2+j))} \Big|_{s=-(2-j)} = A \frac{(-2-j+2-j)}{-(2-j)} + \frac{B(-2-j+2-j)}{-(2-j)+(2+j)} + C$$

$$\frac{5}{(-2+j)(-2+j+2+j)} = C = \frac{5}{(-2+j)(2j)} = \frac{5}{-4j-2} = C \quad B \text{ \& } C \text{ complex conjugates}$$

$$\mathcal{L}^{-1} \left[\frac{5}{s(s+(2-j))(s+(2+j))} \right] = \mathcal{L}^{-1} \left[\frac{1}{s} \right] + \left(\frac{5}{2+4j} \right) \mathcal{L}^{-1} \left[\frac{1}{s+(2+j)} \right] +$$

$$\left(\frac{5}{-2-4j} \right) \mathcal{L}^{-1} \left[\frac{1}{s+(2-j)} \right]$$

Mechanical Solution

$$X(t) = 1 + \left[\frac{5}{-2+4j} \right] e^{-(2+j)t} + \left[\frac{5}{-2-4j} \right] e^{-(2-j)t}$$

$$X(t) = 1 + \left[\frac{5}{-2+4j} \right] e^{-2t-jt} + \left[\frac{5}{-2-4j} \right] e^{-2t+jt}$$

Rationalized denominators

$$\left(\frac{5}{-2+4j} \right) \left(\frac{-2-4j}{-2-4j} \right) = \frac{-10-20j}{4+16} = -\frac{1}{2} - j$$

$$\left[\frac{5}{-2-4j} \right] \left[\frac{-2+4j}{-2+4j} \right] = \frac{-10+20j}{4+16} = -\frac{1}{2} + j$$

$$X(t) = 1 + e^{-2t} \left[\left(-\frac{1}{2} - j \right) e^{-jt} + \left(-\frac{1}{2} + j \right) e^{jt} \right]$$

$$X(t) = 1 + e^{-2t} \left[-\frac{1}{2} e^{-jt} - \frac{1}{2} e^{jt} - j e^{-jt} + j e^{jt} \right]$$

$$X(t) = 1 + e^{-2t} \left[-\frac{1}{2} (e^{-jt} + e^{jt}) + j (e^{jt} - e^{-jt}) \right]$$

$$e^{-jt} + e^{jt} = 2 \cos t \quad \text{Euler's } \quad e^{jt} - e^{-jt} = 2j \sin t \quad \text{Euler's}$$

$$X(t) = 1 + e^{-2t} \left[-\cos t - 2 \sin(t) \right]$$

$$X(t) = 1 - e^{-2t} \left[\cos(t) + 2 \sin(t) \right] \text{ Finally}$$

once again Mathcad can do most of this work

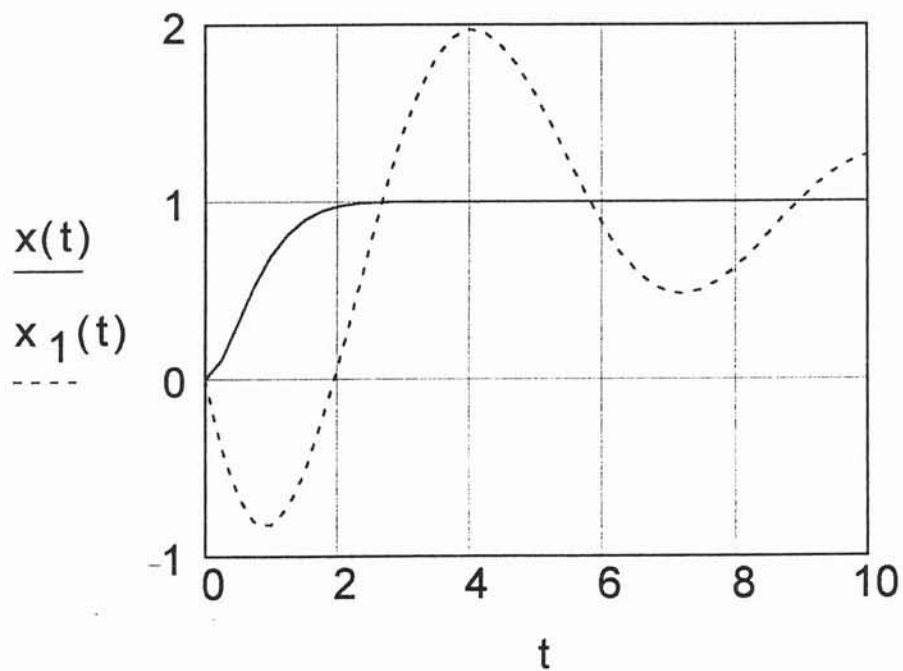
Plot of the mechanical system response

$$x(t) = 1 - e^{-2 \cdot t} \cdot (\cos(t) + 2 \cdot \sin(t))$$

-2 relates to damping of system decrease and see effects

$$x_1(t) = 1 - e^{-2 \cdot t} \cdot (\cos(t) + 2 \cdot \sin(t))$$

$$t = 0, 0.25 \dots 10$$



Transfer Functions

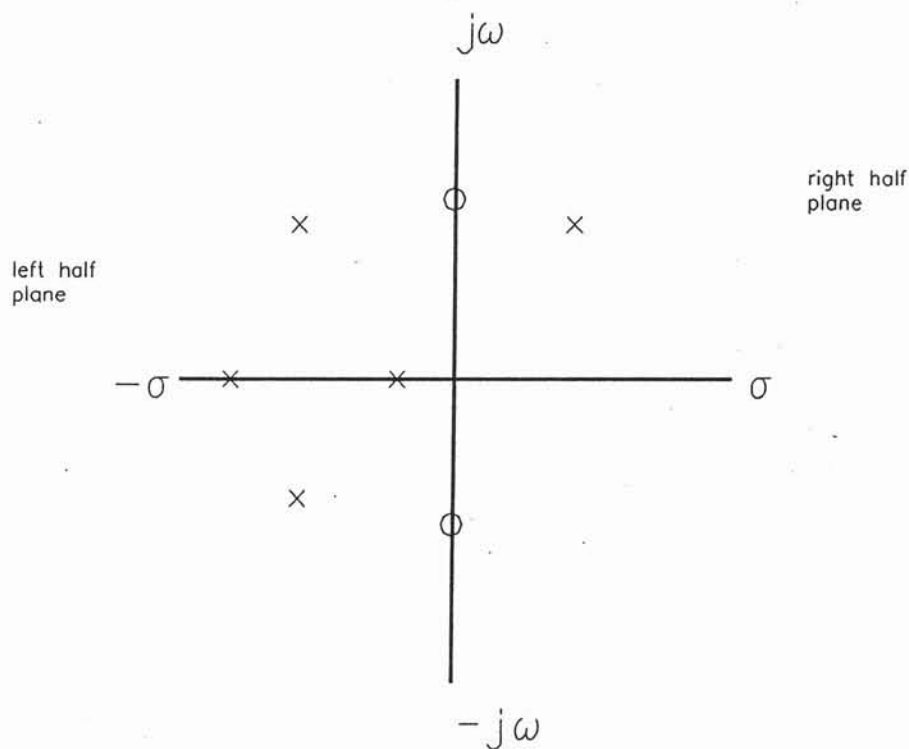
Input/output relationships for a mathematical model usually given by the ratio of two polynomials of the variable s

Definitions

Poles - roots of the denominator polynomial. Values that cause transfer function magnitude to go to infinity.

Zeros - roots of the numerator polynomial. Values that cause the transfer function to go to 0.

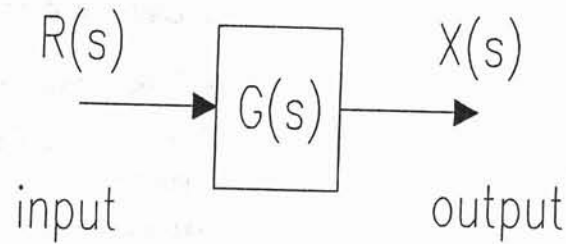
eigenvalues - Characteristic responses of a system. Roots of the denominator polynomial. All eigenvalues must be negative for a system transient (natural response) to decay out.



X's indicate location of pole. O is location of zero

Closer pole is to imaginary axis slower response. Complex roots appear in conjugate pairs

Examples

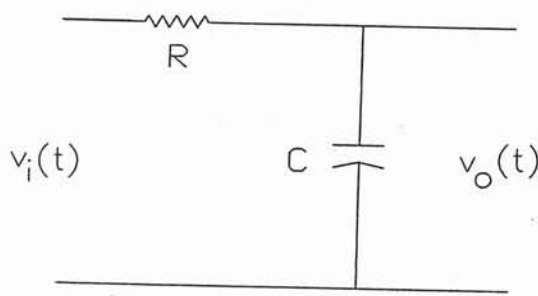


Transfer Function is a "gain" as a function of s

$$X(s) = G(s) \cdot R(s)$$

$$\frac{X(s)}{R(s)} = G(s)$$

Passive Lowpass filter - intergrator



$$V_i(t) = R \cdot i(t) + \frac{1}{C} \int i(t) dt$$

Take Laplace

$$V_i(s) = R \cdot I(s) + \frac{1}{C \cdot s} \cdot I(s)$$

$$\frac{V_i(s)}{R + \frac{1}{C \cdot s}} = I(s)$$

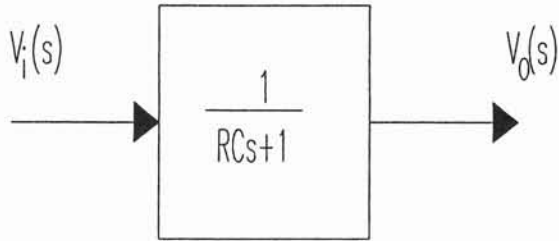
Remember

$$V_o(s) = \frac{1}{C \cdot s} \cdot I(s)$$

So

$$V_o(s) = \frac{\frac{1}{C \cdot s}}{R + \frac{1}{C \cdot s}} \cdot V_i(s) = \frac{1}{R \cdot C \cdot s + 1} \cdot V_i(s)$$

Voltage divider formula



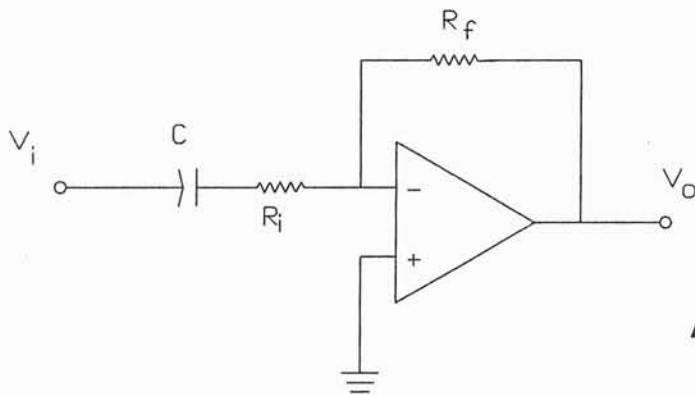
Draw as a block diagram

RC is time constant of system. System has 1 pole at $-1/RC$ and no zeros

Larger RC slower response

Transfer functions of OP AMP circuits

Practical Differentiator- active high pass filter with definite low frequency cutoff.



Take Laplace of components and treat like impedances

General gain formula

$$A_v(s) = \frac{-z_f(s)}{z_i(s)} = \frac{V_o(s)}{V_i(s)}$$

$$z_i(s) = R_i + \frac{1}{C \cdot s} \quad z_f(s) = R_f$$

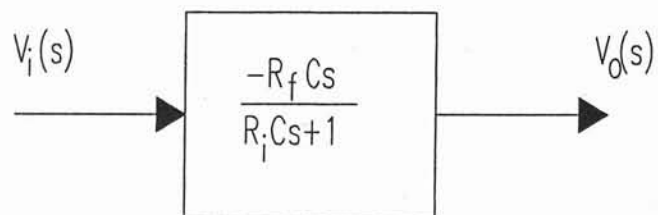
$$A_v(s) = \frac{-R_f}{R_i + \frac{1}{C \cdot s}}$$

Simplify A_v

$$A_v(s) = \frac{-R_f \cdot C \cdot s}{R_i \cdot C \cdot s + 1} = \frac{V_o(s)}{V_i(s)}$$

Transfer function has 1 zero at $s=0$ and 1 pole at $s = -1/R_i C$

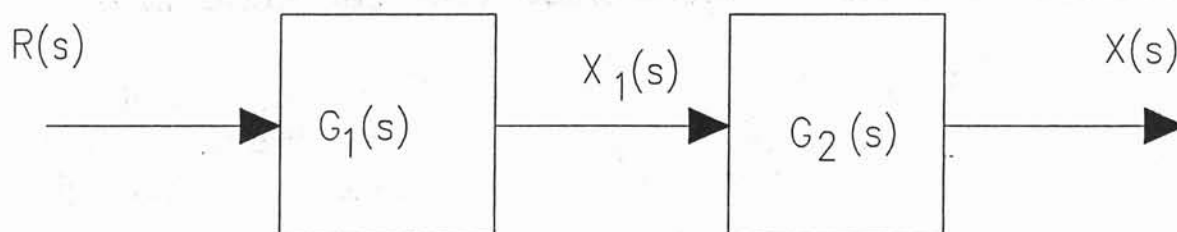
Transfer function for Practical Differentiator



Block Algebra for transfer functions- Cascaded blocks

Series connected - multiply transfer functions

Note: do not cancel common terms from numerator and denominator

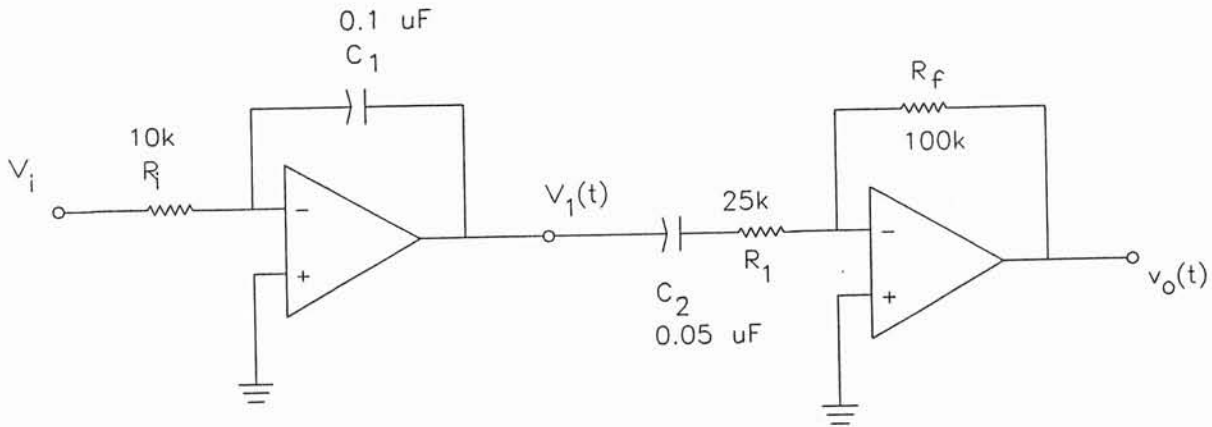


$$X_1(s) = R(s) \cdot G_1(s) \quad (1) \quad X(s) = G_2(s) \cdot X_1(s) \quad (2)$$

$$X(s) = G_1(s) \cdot G_2(s) \cdot R(s) \quad \frac{X(s)}{R(s)} = G_1(s) \cdot G_2(s)$$

Substitute (1) into (2) and simplify to get overall gain

Example with OP AMPs



First stage- integrator Second stage- practical differentiator

$$G_1(s) = \frac{V_1(s)}{V_i(s)} \quad G_2(s) = \frac{V_o(s)}{V_1(s)} \quad \text{Take Laplace of components and use general gain formula}$$

For stage 1

$$A_{v1}(s) = \frac{-z_f(s)}{z_i(s)} \quad z_i(s) = R_i \quad z_f(s) = \frac{1}{C_1 \cdot s} \quad A_{v1}(s) = \frac{-1}{R_i \cdot C_1 \cdot s}$$

Simplify A_{v1} to get $G_1(s)$ $G_1(s) = \frac{-1}{R_i \cdot C_1 \cdot s}$

$G_2(s)$ from previous example

$$A_v(s) = \frac{-R_f \cdot C_2 \cdot s}{R_1 \cdot C_2 \cdot s + 1} = G_2(s)$$

$$\frac{V_o(s)}{V_i(s)} = \frac{-1}{R_i \cdot C_1 \cdot s} \cdot \frac{-R_f \cdot C_2 \cdot s}{R_1 \cdot C_2 \cdot s + 1} \quad \text{Negative signs cancel}$$

$$\frac{V_o(s)}{V_i(s)} = \frac{R_f C_2 s}{(R_1 C_2 s + 1) \cdot (R_i C_1 s)} \quad \text{Simplified form}$$

Plug in given values for the component symbols and compute parameters

$$R_i = 10000 \quad C_1 = 0.1 \cdot 10^{-6} \quad C_2 = 0.05 \cdot 10^{-6}$$

$$R_1 = 25000 \quad R_f = 100000 \quad R_1 \cdot C_2 = 0.001$$

$$R_i \cdot C_1 = 0.001 \quad R_f \cdot C_2 = 0.005$$

Above are all time constants for the system

Final transfer function

$$\frac{V_o(s)}{V_i(s)} = \frac{0.005 \cdot s}{(0.001 \cdot s + 1) \cdot (0.001 \cdot s)}$$

Function has 1 zero at $s = 0$ and two poles $s = -1/0.001 = 1000$ and $s = 0$

Parallel Blocks - add transfer functions

